

(19)



Europäisches Patentamt

European Patent Office

Office européen des brevets



(11)

EP 1 292 043 A1

(12)

## EUROPEAN PATENT APPLICATION

(43) Date of publication:  
12.03.2003 Bulletin 2003/11(51) Int Cl.7: H04B 1/69, H04B 1/713,  
G01S 5/00, H04B 7/185

(21) Application number: 02292173.8

(22) Date of filing: 03.09.2002

(84) Designated Contracting States:  
AT BE BG CH CY CZ DE DK EE ES FI FR GB GR  
IE IT LI LU MC NL PT SE SK TR  
Designated Extension States:  
AL LT LV MK RO SI

(30) Priority: 05.09.2001 GB 0121491

(71) Applicant: Thales Research & Technology (UK)  
Limited  
Addlestone Weybridge, Surrey KT15 2NX (GB)

(72) Inventors:

- Darby, Barry James, Thales Intellectual Property  
94117 Arcueil Cédex (FR)
- Diederich, Pierre, Thales Intellectual Property  
94117 Arcueil Cédex (FR)
- Frazer, Ewan, Thales Intellectual Property  
94117 Arcueil Cédex (FR)

(74) Representative: Chaverneff, Vladimir  
THALES Intellectual Property  
13, Avenue du Président Salvador Allende  
94117 Arcueil Cédex (FR)

## (54) Position fixing system

(57) In a first aspect the present invention provides a method of producing an ultra wide band signal in which (I) a carrier signal is modulated by a spreading waveform which is a periodic deterministic wideband signal and (II) the frequency of the carrier signal is changed by a predetermined amount at predetermined intervals (the reciprocal of this interval being the hopping rate), wherein there is a harmonic relationship between the carrier frequency, the predetermined amount of frequency

change, and the period of the spreading waveform. In this way, an ultra wide band signal is created using both direct sequence and frequency hopping techniques and the signal so created may be detected using a coherent detector at a receiver. This provides a wide bandwidth signal, or even an 'ultra wide bandwidth' (UWB) signal which mitigates the problems of multi-path signals (through coherent combining across a wide bandwidth) and also interference.

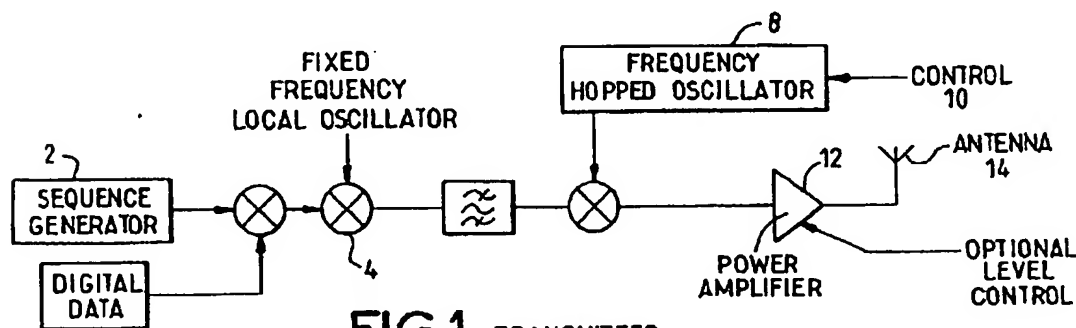


FIG.1 TRANSMITTER

EP 1 292 043 A1

**Description**

[0001] The present invention relates to any positioning system that uses synchronised signals from a number of reference stations. The position of the reference stations is accurately known.

[0002] A mobile terminal receives signals from a number of reference stations, measures the difference in time of arrival between the signals and computes an estimate of its own position relative to the reference stations. In order to form an estimate in three-dimensional space a minimum of four reference station signals must be received.

[0003] A 'dual' arrangement is also well known where each mobile terminal transmits a signal to the reference stations that are mutually time synchronised as before. The reference stations are connected to a central control point where the position calculation is performed. The estimate of position is then transmitted to the mobile terminal. In order to form an estimate in three-dimensional space a minimum of four reference stations must receive the signal from the mobile terminal.

[0004] Factors influencing the accuracy of the position estimate, whichever arrangement is employed, are:

1. Geometry - the direct path to some reference stations may be obscured by natural or man-made objects and/or the mobile can be in a location where trigonometric calculations can give rise to large errors.

2. Multi-path - signals from the reference stations may be subject to reflection off an object (either stationary or moving) and these indirect or multipath signals give rise to an error in the position estimate because they distort the reception of the direct signal.

3. Interference - man-made interference from other systems can block reception of the desired signal because they have much higher power levels or at lower levels of interference can cause errors in the mobile receiver signal processing.

[0005] The GPS system is the best known example of such a positioning system. In this case satellites carry the synchronised reference stations. One of the major deficiencies with GPS is that coverage in buildings and in dense urban environments is not reliable. The reason being the obscuration and attenuation of the direct path signals by buildings. Often in areas where signals may be received they are degraded by complex multi-path components.

[0006] This invention supplements GPS in these environments and can be used where GPS signals are available but greater precision is required e.g. in aircraft landing.

[0007] Spread-spectrum techniques have been de-

veloped to give improvements in a number of instances:

- To enable unambiguous ranging estimates
- To reduce degradation due to multi-path
- To reduce susceptibility to interfering signals
- To reduce detection by unauthorised users
- To reduce interference to other systems

[0008] Often, a number of these improvements can be obtained at the same time. In general, the performance of each of the above is determined directly by the bandwidth and coherence of the spread spectrum signal.

[0009] These techniques are in common use in navigation, communications and radar systems.

[0010] This invention describes a means of producing and receiving a controllable ultra-wideband (UWB) spread spectrum signal that has characteristics that are well matched to the application of precise positioning in environments where GPS is unreliable.

[0011] The nature of the signal also makes it suitable for the other named applications above.

[0012] There are two general methods of generating a spread spectrum signal. The first is by direct modulation of a carrier by a wideband signal and the second is by hopping the carrier frequency.

[0013] In the direct method, the carrier is directly modulated by a deterministic time-limited wideband waveform (spreading waveform). This spreading waveform is made to be periodic by repeating an exact replica after a suitable time interval (waveform period). For the resulting spread spectrum signal to be coherent, the carrier frequency and the waveform period must be harmonically related.

[0014] A common means is to use a pseudo random binary sequence (PRBS) as the spreading waveform. In this case to ensure coherence the carrier frequency and the clock (chipping) rate of the spreading waveform are in a harmonic relationship - this in turn ensures that the time period of the PRBS is harmonically related to the carrier frequency. Another means of producing a coherent spread spectrum signal is to modulate a carrier signal with a chirp waveform that is repeated at a rate that is harmonically related to the carrier frequency. To send messages, the spreading waveform is first modulated with the message signal. The resulting waveform then modulates a carrier.

[0015] In the PRBS example, the chipping rate and the code length of the PRBS signal must be high enough to spread the signal energy over the whole of the desired bandwidth. Each transmission has a unique spreading waveform by which it can be recognised and retrieved and accordingly a receiving station needs to be able to generate the same spreading waveform in synchronism with the transmission in order to recognise and recover the message.

[0016] A multiple access system can be devised by using unique spreading waveforms. All other transmis-

sions, which are combined with different spreading waveforms, appear as noise in the receiver. If the direct modulation is by PRBS the system is termed code division multiple access (CDMA).

[0017] In the frequency hopping method, the frequency of the transmission is changed as a function of a predetermined PRBS. Frequency hopping is usually limited to relatively slow hopping rates (typically a few kilohertz) because of inherent limitations in the equipment used.

[0018] The present invention proposes a method of producing an ultra-wideband (UWB) signal to be produced without some of the limitations of the direct spread spectrum or frequency hopping methods. The present invention combines aspects of these two methods.

[0019] Accordingly in a first aspect the present invention provides a method of producing an ultra wideband signal in which (I) a carrier signal is modulated by a spreading waveform which is a periodic deterministic wideband signal and (II) the frequency of the carrier signal is changed by a predetermined amount at predetermined intervals (the reciprocal of this interval being the hopping rate), wherein there is a harmonic relationship between the carrier frequency, the predetermined amount of frequency change, and the period of the spreading waveform.

[0020] In this way, a coherent UWB signal is created using both direct spread spectrum and frequency hopping techniques and the UWB signal so created is detected using a coherent correlation detector at a receiver.

[0021] A UWB signal is one which has a fractional bandwidth greater than 20% (i.e. 20% of centre carrier frequency). The fractional bandwidth may be defined as  $2(F_H - F_L)/(F_H + F_L)$ , where  $(F_H + F_L)/2$  is the nominal centre carrier frequency, and the majority of the energy lies between lower frequency  $F_L$  and higher frequency  $F_H$ . In this case the energy density outside the  $F_L$  to  $F_H$  region is at least XdB below the largest energy density within the region, the energy density usually being measured in a 1MHz bandwidth. XdB may be 3dB, preferably 10dB or even 20dB. Normally for UWB this will be a bandwidth greater than 500 MHz, and typically between 500 MHz and 7.5GHz.

[0022] Systems with a bandwidth of between 1MHz and 500MHz are usually considered to be Wideband, with narrowband systems having a bandwidth of less than 1MHz, although some put the boundary at 100kHz.

[0023] A deterministic signal is one in which the waveform is substantially or predominantly predictable at a receiver, for example one which does not carry any arbitrary data unknown apriori to the receiver. The deterministic waveform may contain a data pattern known, determined or passed by other means to the receiver. The signal may have deterministic aspects for the part of the waveform used by a receiver to provide position fixing information, i.e. a composite waveform with non-deterministic parts can carry data, whilst the determin-

istic parts carry the position signal. One part of a composite waveform can provide information to make another part or aspect deterministic, for example by carrying forward error correction redundancy. The deterministic waveform can be defined in terms of an explicit mathematical expression of at least one part of its aspects. An example of such an aspect would be the phase of the waveform; in this instance the amplitude need not be deterministic. For two specific embodiments of the present invention (utilising a PRBS signal or a chirp signal respectively) the properties of the deterministic signal to be specified are explained below.

[0024] Preferably the spreading waveform is deterministic across a plurality of frequencies which improves performance of the resolution of finely separated multipath interference. The wider the bandwidth of the plurality of frequencies, the closer the paths which can be separated. The harmonic relationship between the carrier frequency, the frequency change and the spreading waveform period enables information about the transmission channel to be derived by the techniques described.

[0025] Preferably the spreading waveform is a PRBS signal having a determined chipping rate; the code length divided by the chipping rate is the period of the spreading waveform. Alternatively, the spreading waveform may be a chirp signal having a predetermined modulation rate and duration wherein the period of the spreading waveform is the interval between chirps.

[0026] In a further aspect, the present invention provides a method of producing an ultra-wideband (UWB) signal in which (I) a carrier signal is modulated by a PRBS signal having a determined chipping rate and (II) the frequency of the carrier signal is changed by a predetermined amount at predetermined intervals (the reciprocal of this interval is the hopping rate) wherein there is a harmonic relationship between the carrier frequency, the predetermined amount of frequency change, and the chipping rate of the PRBS signal.

[0027] A chirp signal may be used in place of the PRBS signal, and so in a further aspect the present invention provides a method of producing an ultra-wideband signal in which

(I) a carrier signal is modulated by a chirp signal having a predetermined modulation rate and duration and

(II) the frequency of the carrier signal is changed by a predetermined amount at predetermined intervals (the reciprocal of this interval is the hopping rate) wherein there is a harmonic relationship between the carrier frequency, the predetermined amount of frequency change and the interval between chirps.

[0028] The hopping rate may also have a harmonic relationship to the carrier frequency.

[0029] By way of an example, the method could use a PRBS which is a Gold code having code length 1023

and a chipping rate of 1.023MHz. The set of carrier frequencies used in hopping could be at multiples of 1.023MHz and the frequency hopping rate could be 1KHz i.e. the frequency is changed every millisecond or multiples of 1KHz.

Thus, over the period of 1 second, a total sweep of 1.023GHz, can be centered on a carrier of, say, 5GHz. The acquisition of the PRBS code can readily be achieved with GPS-related hardware, and so preferably the code used can be a PRBS of an established or proposed satellite navigation system.

[0030] The formation of a correlation output by coherently summing the results from 1023 successive frequency hopped signals reinforces the desired signal, but tends to cancel or reduce the effect of multipath and interference. As will be appreciated, the signals from each discrete frequency hop are mutually coherent. The more components that can be summed, the better the accuracy.

[0031] Depending on the application the frequency hopping may be carried out in a pseudo-random manner or according to a linear progression. For example, the carrier frequency of each part of the signal may be selected according to a pseudo random algorithm. This improves signal-to-noise characteristics and also, from the point of view of an unauthorised receiver or victim receiver (from the point of view of interference), carrier frequency selection appears random. Over time, the whole available system bandwidth may be used.

[0032] It is also possible for the selection of the carrier frequency (whether done by a pseudo random process or otherwise) to be carried out so as to avoid the use of one or more frequencies or bands of frequencies within the selected bandwidth. This might be done, for example, where it is known that certain portions of the overall bandwidth should be avoided e.g. because it is known that a strong interfering signal might be present or perhaps another service might be compromised (e.g. cellular phones/radar navigation systems, etc.)

[0033] In some embodiments, the UWB signal produced by the method of the present invention may be transmitted discontinuously i.e. on a duty cycle. For example, in a given period of time, the UWB signal may be transmitted for only a portion of that time period (e.g. on for one, off for fifteen etc.). The power of the transmitted signal may be increased during the on portions of the duty cycle with respect to the power which would otherwise be employed if the signal were to be transmitted continuously. Such an arrangement also permits time division multiplexing to be easily arranged for multiple transmitters.

[0034] In addition, the present invention allows a UWB signal to be created which requires a flatter spectrum than can be achieved with a normal UWB signal. A typical UWB signal requires at least a 10 dB allowance at peak to mean across the spectrum, which reduces the power budget for that signal by the same amount. With the present invention, a small headroom (e.g. 2 dB)

should suffice, which in turn permits up to a 50% range increase.

[0035] In preferred embodiments, the spectral shape of the PRBS signal need not follow the usual shape given by the formula  $[(\sin x)/x] \exp 2$ . Various methods for modifying the spectral shape of a PRBS signal are known, for example European Patent EP0131458 discloses a digital method which converts the response of a PRBS into a more rectangular shape using only elementary digital filtering. This technique could be used to configure the set of frequency hops so that any particular frequency segment is avoided, thus avoiding interference to/from a third party as described above.

[0036] Additionally and/or alternatively, in the method of the present invention, power control could be applied to each individual frequency hop in order to control the signal-to-noise ratio of each hop e.g. to achieve the minimum signal-to-noise ratio necessary.

[0037] Similarly, the received signal can be scaled or weighted as a function of the frequency on which it is received in order to improve the accuracy of the time or positional information derived from it. The scaling or weighting function applied may be adapted to improve the accuracy according to the environment experienced by the information passed from the transmitter and receiver. Often the scaling or weighting function will take the form of a "window" [for example, as described in "On the use of Windows for Harmonic Analysis with discrete Fourier Transform", Frederic J. Harris, Proc IEEE, Vol 66, No. 1, January 1978, pp 51-83]. The function may also include low value coefficients for frequency bands where there is perceived to be interference.

[0038] Additionally or alternatively, the signal received in a frequency bandwidth can be replaced using information derived from information received on any or all of the other frequencies, with the intention that this derived information would be a better replica of the transmitted information in that bandwidth, when that information had been lost, corrupted or otherwise modified by the environment experienced by the information passed from the transmitter to the receiver.

[0039] In a further aspect the present invention provides an apparatus for producing an ultra-wideband signal including (I) modulating means for modulating a carrier signal by a spreading waveform which is a periodic deterministic wideband signal and (II) means for changing the frequency of the carrier signal by a predetermined amount at predetermined intervals (the reciprocal of which is the hopping rate), wherein there is a harmonic relationship between the carrier frequency, the predetermined amount of frequency change, and the period of the spreading waveform.

[0040] Preferably the spreading waveform is a chirp signal having a predetermined modulation rate and duration wherein the period of the spreading waveform is the interval between chirps or means for carrying out the frequency hopping in a pseudo-random manner.

[0041] Accordingly, in a further aspect, the present in-

vention provides an apparatus for producing an ultra-wideband signal including (I) modulating means for modulating a carrier signal by a PRBS signal having a determined chipping rate or a chirp signal and (II) means for changing the frequency of the carrier signal by a predetermined amount at predetermined intervals, wherein there is a harmonic relationship between the carrier frequency, the predetermined amount of frequency change, and the chipping rate of the PRBS signal or the interval between chirps.

[0042] In a preferred embodiment, the apparatus includes means for carrying out the frequency hopping in a pseudo-random manner or in any other deterministic pattern

[0043] Preferably the means for changing the frequency of the carrier signal is operable to be carried out so as to avoid the use of one or more frequencies or bands of frequencies within the selected bandwidth.

[0044] The apparatus could be part of a transmitter which may be part of a satellite in a satellite navigation system.

[0045] In a further aspect, the present invention provides an apparatus for receiving a signal including (I) means for demodulating a carrier signal modulated by a spreading waveform which is a periodic deterministic wideband signal and (II) means for detecting the frequency of the carrier which changes by predetermined amounts at predetermined intervals (the reciprocal of which is the hopping rate), and means for changing the carrier frequency to predetermined values.

[0046] In a further aspect, the present invention provides apparatus for receiving an ultra-wideband including (I) means for demodulating a carrier signal modulated by a PRBS signal or a chirp signal and (II) means for detecting the frequency of the carrier which changes signal by a predetermined amount at predetermined intervals, and means for changing the carrier frequency to predetermined values.

[0047] Preferably any receiver is part of or connectable to a satellite navigation signal system receiver.

[0048] Embodiments of the present invention will now be described by way of example with reference to the accompanying drawings in which:

Figure 1 is a schematic diagram of a transmitter according to an aspect of the present invention.

Figure 2 is a schematic diagram of a receiver according to an embodiment of the present invention.

Figure 3 is a graph showing an example of a spectrum of a wide band signal created according to the present invention.

Figure 4 is a graph showing the spectrum of the signal as shown in figure 3, modified to exclude frequencies associated with an interfering signal.

Figure 5 is a schematic diagram showing the frequency hopping used to produce an example of a signal according to the present invention.

Figure 6 is a schematic diagram showing a situation in which the present invention could be used.

Figure 7 shows the dual of the arrangement described in Fig. 6.

Figure 8 shows one example of a frequency hopping sequence for use with the present invention.

[0049] In Figure 1, a sequence generator 2 is used to generate a PRBS signal, this is modulo - 2 added to a data stream and the composite signal then modulates a carrier signal from a local oscillator at mixer 4. Following this, the carrier frequency of the signal is determined by the frequency hopped oscillator 8 under control 10. In one example, control 10 controls the frequency hopped oscillator 8 in a pseudo random fashion so as to produce a signal, part of which is shown in figure 5. The resulting CDMA signal is then transmitted using a standard power amplifier 12 and antenna 14.

[0050] In a preferred embodiment, frequency correction may be applied to one or more of the frequency reference signals e.g. the fixed frequency local oscillator feeding mixer 4 and/or the reference frequency used in frequency hopped oscillator 8. A suitable calibration process may use round trip timings between three or more units in order to calibrate filtered delay. Preferably, the two frequency reference sources in this embodiment (the fixed frequency local oscillator feeding mixer 4 and the frequency reference in frequency hopped oscillator 8) are phase locked i.e. they are really one reference in terms of frequency drift. However, this is just one example of how this could be achieved. Alternatively, the fixed frequency local oscillator and mixer 4 could be omitted and direct synthesis used instead.

[0051] Figure 2 shows a receiver according to an aspect of the present invention. The CDMA signal according to the present invention is received using a standard antenna 20 and low noise amplifier 22. The received signal will be a frequency hopped signal and the frequency hopping aspect is effectively removed using demodulator 24. Demodulator 24 uses a frequency hopped oscillator 26 under control of a synchronisation unit 28 which is operating in accordance with the same algorithm used to produce the frequency hopping in the transmitter.

[0052] In this way, the signal produced at an input terminal 30 of a GPS receiver 32 is effectively a standard CDMA signal as if it had been produced by the direct sequence method only and having a single fixed carrier frequency. The GPS receiver 32 is then able to decode the signal using the same PRBS code which was used to produce the signal, in the known decoding manner.

[0053] In a more general embodiment the GPS receiver can be replaced by a processor allowing weight-

ing to be applied to different frequencies.

[0054] Preferably the receiver includes means for resilient recovery of frequency error without the need for additional parallel hardware; this improves the operational range of the equipment. The means includes processing means for processing the input signal in accordance with an algorithm which operates as follows:

[0055] Firstly, the integrated complex number vectors for each frequency which make up the input terms of the Fourier Transform is determined. Before each particular frequency, the complex number vectors for its use on one occasion and a subsequent occasion are determined and one is multiplied by the complex conjugate of the other - the result gives the phase change over the time interval between the visits to that particular frequency. The same may then be done for other frequencies used in the frequency hop system. In some embodiments, the phase changes may be scaled to correct for any differences in the respective time intervals. If all the time intervals are equal (e.g. because the same repeating hop sequence is used to visit all frequencies in the chosen set) then no correction is required. The set of (possibly corrected) vectors may then be put through a windowed Fourier Transform using one or more window functions. The peak term in the Fourier Transform output represents a frequency area in the transmission, including any Doppler error produced as a result of movement. The input terms to the Fourier Transform may (instead of or as well as the above) be filtered in other ways, e.g. the phase locked loop, Kalman filter etc. in order to track the frequency error.

[0056] This method may provide several advantages as follows:-

- a. It has processing gain and hence copes with large levels of noise.
- b. It is not degraded by (stationary) multipath, indeed it benefits from it.
- c. It copes with large frequency offsets with no ambiguity. Cheap uncorrected crystals can be used without compromising range.
- d. It does not require anymore high speed hardware to implement.
- e. The algorithm works with any set of time spacings. Larger time spacings between visits to a frequency result in more accurate measurements of frequency error (but less range - it may need a coarse correction first). One would therefore start with adjacent visits for coarse correction, then track more finely using visits several visits apart. Coarse correction can be done by adjusting the frequency reference, or by multiplying the terms prior to the Fourier transform by a vector rotating (backwards) at the rate of coarse estimate. The transmit pattern can be continuous, or it can be pulsed, provided at least two visit cycles complete adjacent to one another for the first coarse estimate. It is well suited to time division multiplexing.

f. The benefit is primarily of use to positioning, and low rate data systems, and not to high rate data systems.

g. A side effect of the Fourier transform is a segment of the impulse response of the channel is also established (about 30 metres is typical). Pulse systems would have to have hundreds of Rake taps to do the same (which would be very current consuming), or would have to scan the channel sequentially (which either reduces the update rate, or reduces the integration period and hence the range). It should be possible to track movement at rates about 100x faster compared to pulse UWB at the same range.

[0057] Figure 3 shows a graph of frequency against amplitude for an example of a CDMA signal produced in accordance with the present invention. The signal is referred to as a hybrid frequency hopping/direct sequence spread spectrum signal. As can be seen, the spectrum of the signal consists of a plurality of individual signal peaks (one of which is labelled 40). Each peak 40 is centred around a different carrier frequency and the plurality of peaks - which may overlap - combine to produce a hybrid signal covering the wide bandwidth shown. For example, a bandwidth of 1GHz would allow positioning accuracies better than 0.2m and multipath resolution of better than 0.5m.

[0058] Figure 4 is another graph of amplitude against frequency for an example of a signal according to the present invention. In this example, the frequency hopping has been modified so as to exclude a band of frequencies shown as 'x'. This band of frequencies corresponds with an expected interfering signal 42, so that the interfering signal does not interfere with the signal of the present invention, and vice versa.

[0059] Figure 5 is a schematic graph of frequency against time for part of a signal in accordance with the present invention. As can be seen from Figure 3, the aim of the invention in this example is to produce a signal having a bandwidth from frequency F1 to frequency F2. In Figure 5, from time T1 to time T2 a 'burst' of signal 50 is produced which, by the PRBS method explained previously, produces a signal having a bandwidth from frequency F3 to frequency F4. At time T2, the carrier frequency has changed and from time T2 to T3 a further burst of signal 52 is produced, again by the PRBS method, having a bandwidth from frequency F5 to frequency F6.

[0060] As can be seen from Figure 5, this procedure continues and effectively discrete bursts of signal are produced covering a different part of the system bandwidth which, when added together, may cover all, substantially all or part of the system bandwidth. As explained previously, part of the system bandwidth may remain deliberately unused and in this example that is from frequency F7 to frequency F8.

[0061] The bandwidth can usefully range from 1 MHz

(1.023 MHz is convenient for GPS) through say 10.23MHz (military GPS chipping rate) to say 100MHz (here the acquisition problem is paramount) A good example would be in the range 10-20MHz. Acquisition times are governed by bandwidth, sequence period and carrier frequency offset.

[0062] Frequency hopping may or may not take place in a linear fashion. In Fig. 5 hopping is arranged in a pseudo random manner so that each signal burst occupies an apparently unrelated part of the system bandwidth to the previous burst.

[0063] In summary, Figure 5 shows in diagrammatic form the transmitted waveform for this Ultra-wideband scheme. It is essentially a hybrid of two methods of spread-spectrum namely frequency hopping (FH) and, for example, direct sequence (PRBS) spread-spectrum (DSSS). The available system bandwidth is regarded as being sub-divided into a number (N) of contiguous channels each of bandwidth B. The direct sequence signal (spreading waveform) occupies a notional bandwidth B and has a period T (T may be the duration of one or more code cycles). Successive pulses of the DSSS signal are transmitted on carrier frequencies determined by an algorithm known by both transmitter and receiver, the receiver performs a process known as acquisition to enable signal reception. From the point of view of an unauthorised receiver, carrier frequency selection appears random. Over time, the whole available system bandwidth may be used. By correctly combining all of the signals at the receiver it is possible to gain the benefit of the very wide spread in bandwidth  $N \times B$  (or less if overlapped) to discriminate against multi-path signals. In some embodiments there will be an advantage in combining only some of the signals and accepting a reduced bandwidth, for example if some signals are affected by noise or interference. Figure 5 also shows a channel (it could be more than one, either in adjacent slots or in a number of non-adjacent slots) that is (are) not available for use in order to prevent or avoid interference.

[0064] In order to take advantage of the processing gain enabled by the ultra wide bandwidth, the frequency hopping is preferably phase coherent (constant phase different is maintained between transmitter and receiver as hopping progresses). This is achieved by generating carrier frequencies that are harmonic multiples of a fundamental frequency.

[0065] This invention may use the established signal structures used by GPS (and in future could be based on any such standard e.g. as being developed for GALILEO) that ensures that channels are spaced by that value (1.023 MHz). This in principle enables standard GPS correlation receivers to be used.

[0066] In Figure 6, a receiver 60 is shown which can receive signals from four reference transmitters 62. The reference transmitters have known locations and are time synchronised. The receiver 60 can receive direct path transmissions from the transmitter 62 and one example is shown as path 64. However the receiver 60

may also receive indirect path transmissions, such as reflected transmissions from a reflecting object 66. Two examples of indirect path transmissions are shown as 68, 70.

[0067] The reception of the indirect path transmissions produces a multipath echo and since these indirect path transmissions are slightly delayed this potentially shifts the position estimate for the receiver. The ultra wide band signals used by the present invention helps to allow the receiver to detect smaller time differences, thereby reducing the position errors due to the multipath nature of the transmissions.

[0068] Figure 7 depicts a dual of the arrangement described by Figure 6. Receivers at the fixed reference station 72 pick up the transmission of one (or more) mobile station 70, 71. As in Figure 6, the reference stations are accurately synchronised and the position of the mobile station can be calculated from the time differences of the signals observed at the reference stations. For a position in n dimensions, as in Figure 6, at least (n + 1) fixed stations are required. The received signals may be direct 74 or multipath 80.

[0069] Figure 8 shows one example of a frequency hopping sequence used with the present invention. It has the following features:-

- a. Each frequency is visited once.
- b. The synthesiser doesn't have to change far in frequency from one hop to the next. This makes fast hopping easier.
- c. The arrangement allows for multiple transmitters which "follow" each other. I.e. when one transmitter finishes on a frequency, the next one then uses that frequency. This provides a time-division/frequency division multiplexing scheme.
- d. The multiplex scheme is arranged so that multiple transmitters are spaced apart in frequency to avoid interference and near-far issues.
- e. The scheme is represented by equations, allowing control of frequency spacing, number of transmitter slots, and total number of frequencies (2 of these 3) to be controlled.

[0070] The following is an example set of equations to derive a suitable frequency hopped sequence.

Let W be the number of cycles (e.g. 8); determines the number of frequencies and channels. W is even when  $Z=1$ .

Let S be the frequency step (e.g. 7), e.g. for near far spacing.

$S \times R$  is the number of frequencies (when  $Z=1$ )

$W^2$  is the number of channels.

Let  $R = \text{INT}(S/2)+1$

Let  $Y = -1$  typically, the stagger.

Let  $Z = 0$  or  $1$  (frequency spacing doubler)

Let  $X = \text{IF}(Z, 0.5, 0)$

Let  $A$  be  $0..S^*R-1$ , as increasing time ( $x$  axis)

$B = \text{INT}(A/W)$

$C = B^{\circ} \cdot (S^*2)$

$D = C^{\circ} \cdot 2$

$E = D^*R$

$F = \text{INT}((C^{\circ} \cdot (S^*2))/S)$

$G = F^*S$

$H = \text{INT}(C/2)$

$I = G - H + 2(1-F)^*H - F + D^*F^*Y$

$J = I + E$

$K = A^{\circ} \cdot W$

$L = \text{INT}(A/W)^{\circ} \cdot 2$

$M = L^*S^*(W-1)$

$N = M - K^*S + 2(1-L)^*K^*S$        $P = J + N + \text{IF}((A\%$   
 $(W^*S^2)) > -(W^*S).X.0) * \text{IF}(Z, 2, 1)$

$P$  is the frequency number.

$\text{IF}(A, B, C)$  means if  $A$  is true (non zero), then  $B$  else  $C$ .

$\text{INT}(X)$  means the integer part of  $X$ .

$\%$  is the modulus operator.

$A-Z$  are variables.

**[0071]** A family of sequences are found by adjusting integers  $W$ ,  $S$  and by setting  $Z$  as  $0$  or  $1$ .  $Y$  may also be adjusted.

**[0072]** Additionally or alternatively, a system can be formed where a mobile station is not required in the embodiment, but in which the reference stations monitor their own and/or each others transmissions for changes to the multipath signal. The range or pseudorange at which these signals are seen can be determined by the same method as if they were received by a mobile station apparatus. By triangulation of the ranges or pseudoranges seen it is possible to determine the position and/or movement of one or more unequipped mobile targets within the range of the reference stations. Where each reference station monitors its own signal, it is referred to as a monostatic radar.

**[0073]** A tag system is also possible with a single transmitter and receiver, or a transponder pair. Where each reference station monitors another reference station it is a bistatic radar. Where each reference stations

monitor two or more reference stations (one of which could be its own signal), it is a multistatic radar. The reference stations may also monitor a reduced number of transmitted signals.

**[0074]** The method can also be extended to one or more mobile reference stations provided they calculate their relative positions. In each case, the range(s) or pseudorange(s) to the perturbation(s) in the multipath signal are known, and from these the position of one or more moving targets can readily be determined, and they can be tracked. A system designed to provide location information to an equipped mobile will necessarily also provide location information (2D or 3D) when used to locate unequipped mobiles. Applications for this technique include but are not limited to intrusion sensors and safety warning systems such as collision warnings. An alarm can be raised if one or more target(s) are present in particular areas or volumes.

**[0075]** Filtering can be applied by determining, for example, the speed or acceleration of the target using the received signals in order to reduce false alarms or predict collisions. The gain or sensitivity can also be adjusted according to range or degree of multipath, for example. Alarms may be suppressed where a mobile apparatus whose position has been determined has informed the reference stations, and its location is at or close to the perturbation. For systems that aid safety, machinery may be shut down, or vehicles slowed or stopped, and/or an alarm sounded if an area is entered, for example.

**[0076]** The embodiments described here are particularly suited to radar applications, since they do not require additional hardware to scan for single or multiple targets, unlike current instances in the state of the art of UWB. The embodiments also have significant advantages in their ability to separate and classify multiple targets due to the fine resolution of multipath reflection that they can determine in the presence of other signals or clutter. The embodiments may be combined with other positioning or timing systems such as GPS to provide greater coverage, availability and/or accuracy.

**[0077]** It will be appreciated that these embodiments of the invention are given by way of example only and modifications will be apparent to those skilled in the art.

## Claims

1. A method of producing an ultra-wideband signal in which (I) a carrier signal is modulated by a spreading waveform which is a periodic deterministic wide-band signal and (II) the frequency of the carrier signal is changed by a predetermined amount at predetermined intervals (the reciprocal of this interval being the hopping rate), wherein there is a harmonic relationship between the carrier frequency, the predetermined amount of frequency change, and the period of the spreading waveform.



2. A method according to claim 1 in which the spreading waveform is a PRBS signal having a determined chipping rate, the code length divided by the chipping rate is the period of the spreading waveform.
3. A method according to claim 1 in which the spreading waveform is a chirp signal having a predetermined modulation rate and duration wherein the period of the spreading waveform is the interval between chirps.
4. A method according to any of the above claims wherein the frequency hopping is carried out in a pseudo-random manner.
5. A method according to any of the above claims wherein the carrier frequency of each part of the signal may be selected according to a pseudo-random algorithm.
6. A method according to any of the above claims wherein the selection of the set of carrier frequencies is carried out so as to avoid the use of one or more frequencies or bands of frequencies within the selected bandwidth.
7. A method according to any of the above claims wherein power control is applied to each individual frequency hop in order to control the signal-to-noise ratio of each hop or to reduce interference to third parties.
8. A method according to any of the above claims wherein the information received is weighted or scaled according to a function of the frequency on which it is received.
9. A method according to any of the above claims in which the information received on a frequency is replaced with information derived from the information received on any of the frequencies prior to determining positional or time information.
10. A method according to claim 9 in which the replaced information is an estimation of the information that would have been received if the environment experienced by the information passed from the transmitter and receiver was more favourable.
11. An apparatus for producing an ultra-wideband signal including (I) modulating means for modulating a carrier signal by a spreading waveform which is a periodic deterministic wideband signal and (II) means for changing the frequency of the carrier signal by a predetermined amount at predetermined intervals (the reciprocal of which is the hopping rate), wherein there is a harmonic relationship between the carrier frequency, the predetermined

amount of frequency change, and the period of the spreading waveform.

12. Apparatus according to claim 11 in which the spreading waveform is a PRBS signal having a determined chipping rate, the code length divided by the chipping rate is the period of the spreading waveform.
13. Apparatus according to claim 11 in which the spreading waveform is a chirp signal having a predetermined modulation rate and duration wherein the period of the spreading waveform is the interval between chirps.
14. Apparatus according to claim 11, 12 or 13 including means for carrying out the frequency hopping in a pseudo-random manner.
15. Apparatus according to claim 14 wherein the means for changing the frequency of the carrier signal is operable to be carried out so as to avoid the use of one or more frequencies or bands of frequencies within the selected bandwidth.
16. A transmitter including apparatus according to any of claims 11-15.
17. A positioning system including a plurality of reference stations, each reference station including a transmitter according to claim 16.
18. A positioning system according to claim 17 wherein the reference stations are part of a cellular telephone network.
19. A positioning system according to claim 17 or 18 including a mobile terminal which includes apparatus according to claim 21.
20. A positioning system including a mobile terminal which includes a transmitter according to claim 16, the positioning system also including a plurality of reference stations which each include apparatus according to 21.
21. Apparatus for receiving a signal including (I) means for demodulating a carrier signal modulated by a spreading waveform which is a periodic deterministic wideband signal and (II) means for synchronising a local oscillator to match the frequency of the carrier which changes by predetermined amounts at predetermined intervals (the reciprocal of which is the hopping rate), and means for changing the carrier frequency to predetermined values.
22. Apparatus according to claim 21 in which the spreading waveform is a PRBS signal having a de-

terminated chipping rate, the code length divided by the chipping rate is the period of the spreading waveform.

23. Apparatus according to claim 21 in which the spreading waveform is a chirp signal having a predetermined modulation rate and duration wherein the period of the spreading waveform is the interval between chirps. 5
24. A method of producing an ultra-wideband signal substantially as any of the embodiments herein described with reference to the accompanying drawings. 10
25. An apparatus for producing an ultra-wideband signal substantially as any of the embodiments herein described with reference to the accompanying drawings. 15
26. A positioning system substantially as any of the embodiments herein described with reference to the accompanying drawings. 20
27. Apparatus for receiving a signal substantially as any of the embodiments herein described with reference to the accompanying drawings. 25

30

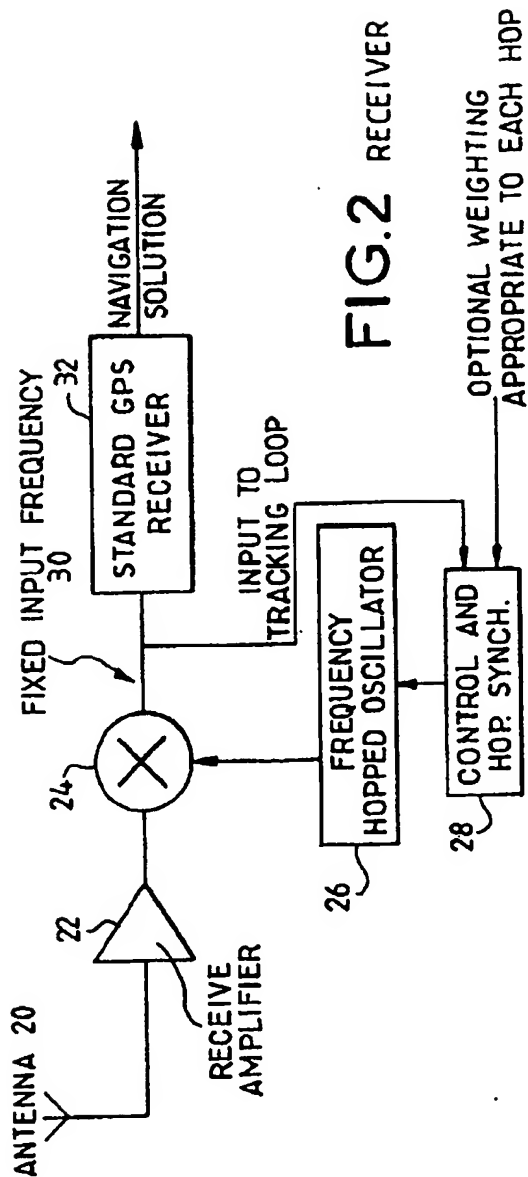
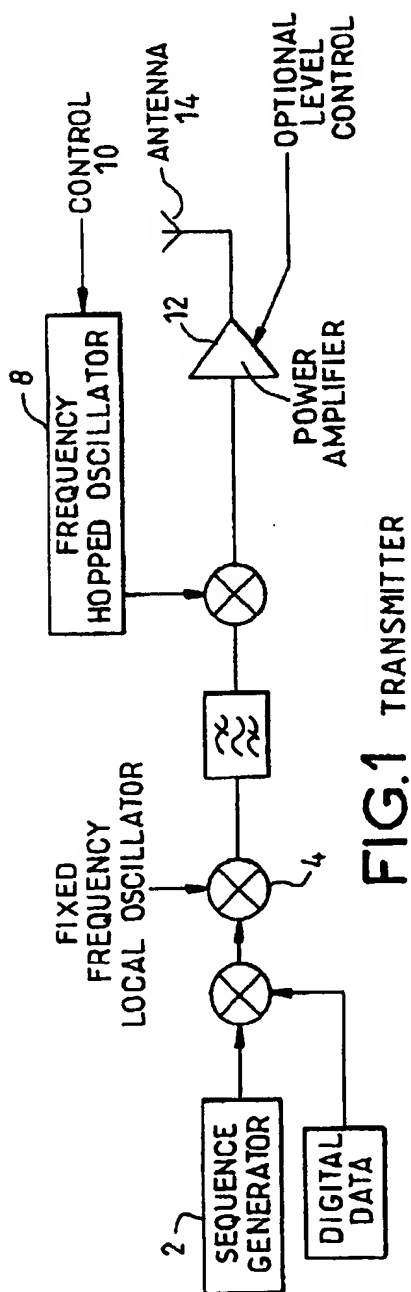
35

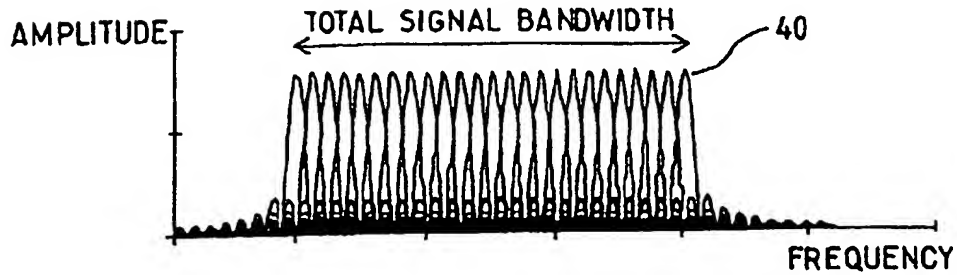
40

45

50

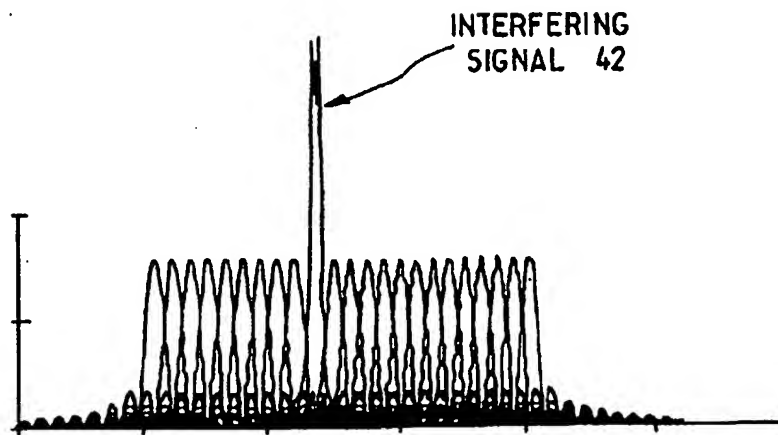
55





SPECTRUM OF HYBRID FREQUENCY HOPPING-DIRECT SEQUENCE  
SPREAD SPECTRUM SIGNAL

FIG.3



SPECTRUM OF SIGNAL ADAPTED TO EXCLUDE THE USE OF FREQUENCIES  
IN THE SAME BAND AS A STRONG INTERFERING SIGNAL

FIG.4

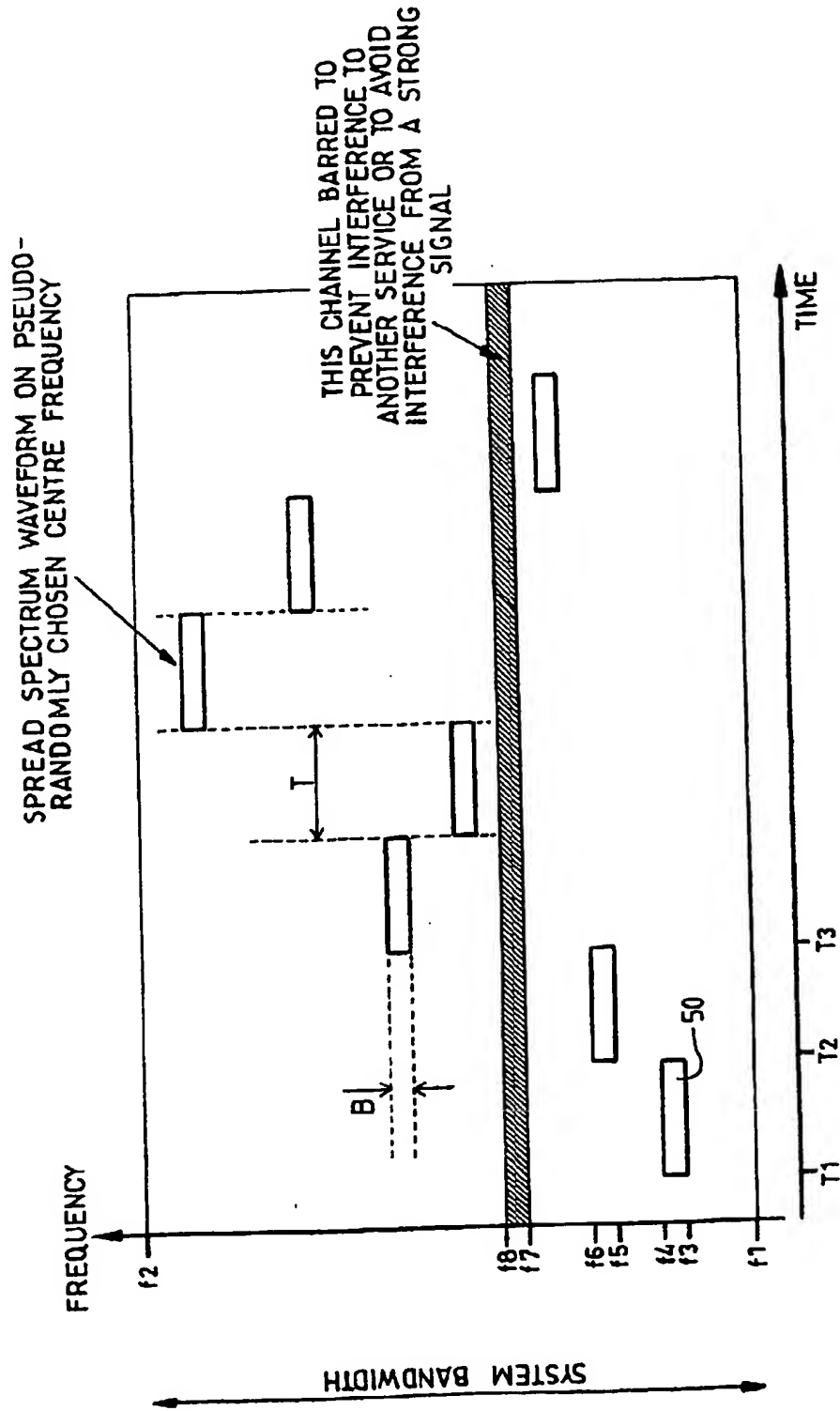
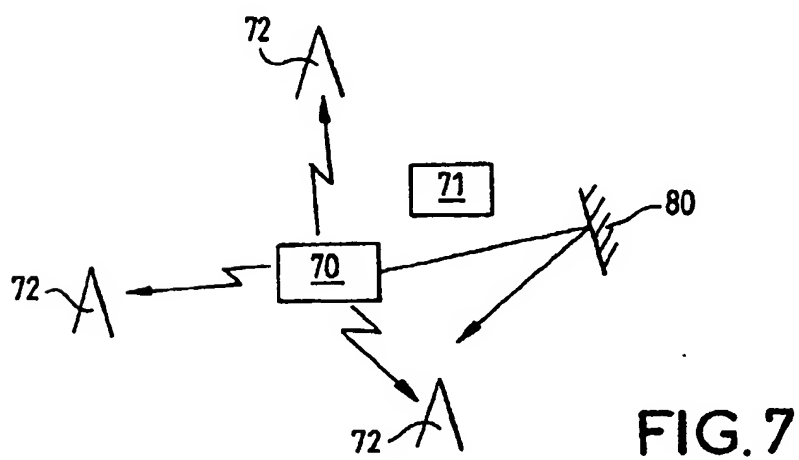
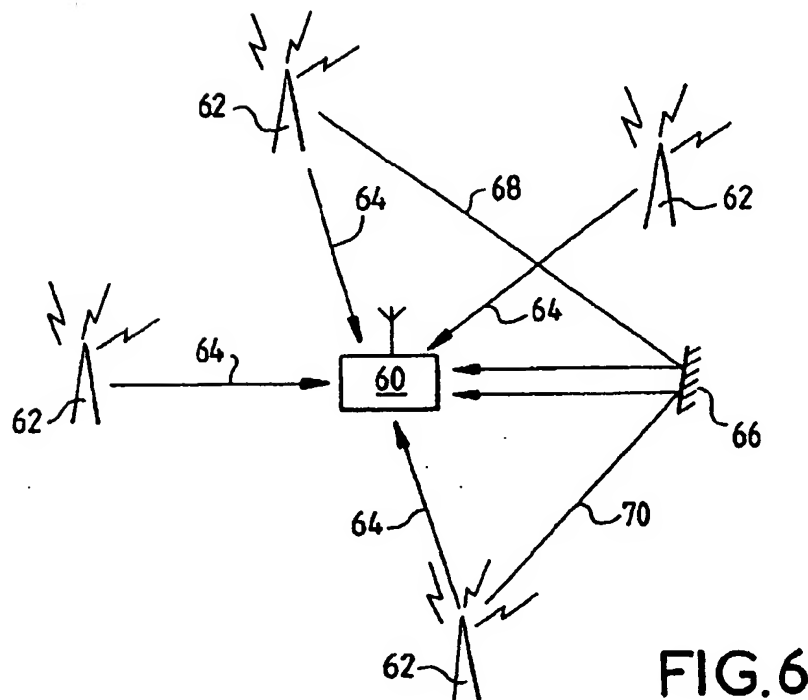


FIG. 5



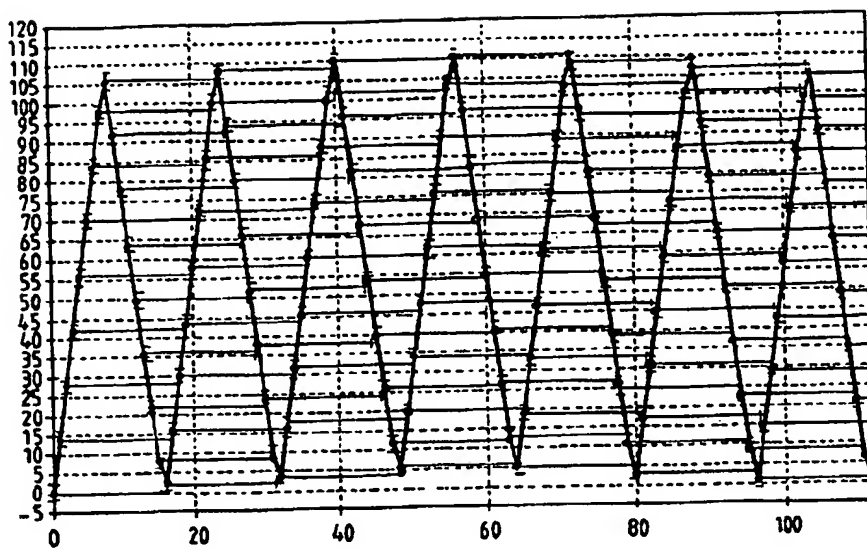


FIG.8



European Patent  
Office

# PARTIAL EUROPEAN SEARCH REPORT

Application Number

which under Rule 45 of the European Patent Convention EP 02 29 2173  
shall be considered, for the purposes of subsequent  
proceedings, as the European search report

DOCUMENTS CONSIDERED TO BE RELEVANT			
Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int.Cl.7)
X	WO 00 03502 A (QUALCOMM INC) 20 January 2000 (2000-01-20) * abstract; figures 1,2A,3,4 * * page 7, line 11 - line 14 * * page 9, line 2 - line 5 * ---	1-23	H04B1/69 H04B1/713 G01S5/00 H04B7/185
X	US 5 706 275 A (ZHENGDI QIN) 6 January 1998 (1998-01-06) * abstract; claims 1-8; figures 3,4A,4B * * column 3, line 45 - column 4, line 12 * ---	1-23	
Y	US 5 222 098 A (HOSHIKUKI ATSUSHI ET AL) 22 June 1993 (1993-06-22) * abstract; figures 1,4 * * column 1, line 37 - column 2, line 29 * * column 4, line 25 - line 57 * ---	1-23	
Y	US 6 215 810 B1 (PARK CHAN-BUM) 10 April 2001 (2001-04-10) * abstract; figures 3,4 * * column 4, line 28 - line 62 * * column 5, line 27 - column 7, line 8 * --- -/--	1-23	
			TECHNICAL FIELDS SEARCHED (Int.Cl.7)
			H04B G01S
<b>INCOMPLETE SEARCH</b>			
<p>The Search Division considers that the present application, or one or more of its claims, does/does not comply with the EPC to such an extent that a meaningful search into the state of the art cannot be carried out, or can only be carried out partially, for these claims.</p> <p>Claims searched completely:</p> <p>Claims searched incompletely:</p> <p>Claims not searched:</p> <p>Reason for the limitation of the search:</p> <p>see sheet C</p>			
Place of search		Date of completion of the search	Examiner
MUNICH		11 November 2002	Ricciardi, M
<b>CATEGORY OF CITED DOCUMENTS</b> X : particularly relevant if taken alone Y : particularly relevant if combined with another document of the same category A : technological background O : non-written disclosure P : intermediate document T : theory or principle underlying the invention E : earlier patent document, but published on, or after the filing date D : document cited in the application L : document cited for other reasons & : member of the same patent family, corresponding document			

EPO FORM 150 (6.82) (9-00/07)





Europäisches  
Patentamt

UNVOLLSTÄNDIGE RECHERCHE  
ERGÄNZUNGSBLATT C

Nummer der Anmeldung

EP 02 29 2173

Claim(s) searched completely:  
1-23

Claim(s) not searched:  
24-27

Reason for the limitation of the search:

According to Rule 29(6) EPC, the claims shall not rely on references to drawings, except where absolutely necessary. The generic formulation of claims 24-27 can be related to an extremely large number of possible apparatuses and methods, since no characterising features are directly and unambiguously ascertainable from the drawings. In fact, the claims contain so many options, variables, possible permutations and provisos that a lack of clarity (and conciseness) within the meaning of Article 84 EPC arises to such an extent as to render a meaningful search of the claims impossible.

It is here recalled that the onus is upon the applicant to show that it is "absolutely necessary" to rely on reference to the drawings for said claims 23-27 (cf. T150/82, OJ 7/1984)



European Patent  
Office

## PARTIAL EUROPEAN SEARCH REPORT

Application Number  
EP 02 29 2173

DOCUMENTS CONSIDERED TO BE RELEVANT			CLASSIFICATION OF THE APPLICATION (Int.Cl.7)
Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	
A	US 5 701 328 A (SCHUCHMAN LEONARD ET AL) 23 December 1997 (1997-12-23) * the whole document *	1-23	
A	GLAS J P F: "On multiple access interference in a DS/FFH spread spectrum communication system" SPREAD SPECTRUM TECHNIQUES AND APPLICATIONS, 1994. IEEE ISSSTA '94., IEEE THIRD INTERNATIONAL SYMPOSIUM ON OULU, FINLAND 4-6 JULY 1994. NEW YORK, NY, USA, IEEE, 4 July 1994 (1994-07-04), pages 223-227, XP010129687 ISBN: 0-7803-1750-5 * abstract; figure 1 * * page 223, left-hand column, paragraph 4 - right-hand column, paragraph 2 *	1-23	
A	US 5 596 330 A (MEIMAN YEHOUDA ET AL) 21 January 1997 (1997-01-21) * abstract * * column 13, line 24 - column 15, line 19 *	1-23	
			TECHNICAL FIELDS SEARCHED (Int.Cl.7)

EPO FORM 1503 (04/97) (P.O. 10)

**ANNEX TO THE EUROPEAN SEARCH REPORT  
ON EUROPEAN PATENT APPLICATION NO.**

EP 02 29 2173

This annex lists the patent family members relating to the patent documents cited in the above-mentioned European search report. The members are as contained in the European Patent Office EDP file on the European Patent Office is in no way liable for these particulars which are merely given for the purpose of information.

11-11-2002

Patent document cited in search report		Publication date	Patent family member(s)	Publication date
WO 0003502	A	20-01-2000	AU 5096799 A	01-02-2000
			CN 1308796 T	15-08-2001
			EP 1213856 A1	12-06-2002
			EP 1097528 A1	09-05-2001
			JP 2002520945 T	09-07-2002
			TW 469709 B	21-12-2001
			WO 0003502 A1	20-01-2000
US 5706275	A	06-01-1998	FI 945630 A	30-05-1996
			AT 223124 T	15-09-2002
			AU 700703 B2	14-01-1999
			AU 3985395 A	19-06-1996
			CN 1140000 A ,B	08-01-1997
			DE 69527944 D1	02-10-2002
			EP 0745291 A1	04-12-1996
			WO 9617446 A1	06-06-1996
			JP 9509029 T	09-09-1997
			NO 963144 A	26-09-1996
US 5222098	A	22-06-1993	JP 2082839 C	23-08-1996
			JP 4047726 A	17-02-1992
			JP 7118668 B	18-12-1995
US 6215810	B1	10-04-2001	AU 712667 B2	11-11-1999
			AU 1317199 A	12-08-1999
			CN 1232330 A	20-10-1999
			GB 2335830 A ,B	29-09-1999
			JP 3193012 B2	30-07-2001
			JP 11261439 A	24-09-1999
US 5701328	A	23-12-1997	US 5604765 A	18-02-1997
			AU 5354796 A	11-09-1996
			WO 9626591 A1	29-08-1996
			AU 4684596 A	19-07-1996
			WO 9620542 A1	04-07-1996
			US 6111538 A	29-08-2000
			US 6160837 A	12-12-2000
US 5596330	A	21-01-1997	US 5592180 A	07-01-1997
			US 5583517 A	10-12-1996
			US 5430759 A	04-07-1995
			US 5335246 A	02-08-1994
			US 5379047 A	03-01-1995
			EP 0813689 A1	29-12-1997
			WO 9625673 A1	22-08-1996
			AU 8114194 A	13-06-1995

EPO FORM P456

For more details about this annex : see Official Journal of the European Patent Office, No. 12/82

**ANNEX TO THE EUROPEAN SEARCH REPORT  
ON EUROPEAN PATENT APPLICATION NO.**

EP 02 29 2173

This annex lists the patent family members relating to the patent documents cited in the above-mentioned European search report.  
The members are as contained in the European Patent Office EDP file on  
The European Patent Office is in no way liable for these particulars which are merely given for the purpose of information.

11-11-2002

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
US 5596330 A	WO	9514936 A1	01-06-1995
	AU	8114094 A	13-06-1995
	WO	9514935 A1	01-06-1995
	AU	8113994 A	13-06-1995
	EP	0730808 A1	11-09-1996
	WO	9515064 A2	01-06-1995
	SG	45367 A1	16-01-1998
	US	5546422 A	13-08-1996
	US	5870426 A	09-02-1999
	US	5499266 A	12-03-1996
	US	5519718 A	21-05-1996
	US	5712867 A	27-01-1998
	AT	160882 T	15-12-1997
	AU	4476093 A	24-02-1994
	AU	4476293 A	03-03-1994
	CA	2104544 A1	21-02-1994
	CA	2104555 A1	21-02-1994
	DE	69203485 U1	15-01-1998
	DE	69203485 T2	23-04-1998
	EP	0583523 A2	23-02-1994
	EP	0583523 A2	23-02-1994
	IL	104264 A	23-07-1996
	IL	104265 A	15-04-1997
	US	5530452 A	25-06-1996
-----			

EPO FORM P459

For more details about this annex : see Official Journal of the European Patent Office, No. 12/82

**This Page is Inserted by IFW Indexing and Scanning  
Operations and is not part of the Official Record**

## **BEST AVAILABLE IMAGES**

Defective images within this document are accurate representations of the original documents submitted by the applicant.

Defects in the images include but are not limited to the items checked:

☒ **BLACK BORDERS**

☐ **IMAGE CUT OFF AT TOP, BOTTOM OR SIDES**

☐ **FADED TEXT OR DRAWING**

☐ **BLURRED OR ILLEGIBLE TEXT OR DRAWING**

☐ **SKEWED/SLANTED IMAGES**

☐ **COLOR OR BLACK AND WHITE PHOTOGRAPHS**

☐ **GRAY SCALE DOCUMENTS**

☐ **LINES OR MARKS ON ORIGINAL DOCUMENT**

☐ **REFERENCE(S) OR EXHIBIT(S) SUBMITTED ARE POOR QUALITY**

☐ **OTHER:** \_\_\_\_\_

**IMAGES ARE BEST AVAILABLE COPY.**

**As rescanning these documents will not correct the image problems checked, please do not report these problems to the IFW Image Problem Mailbox.**